

## SPECIFICATION

**An Automatic Zero Point Correction Device for a Pressure Sensor,  
a Pressure Control Device and a Pressure Type Flow Rate Control  
Device**

This is a National Phase Application in the United States of International Patent Application No. PCT/JP2004/008118 filed June 10, 2004, which claims priority on Japanese Patent Application No. 2003-177135, filed June 20, 2003. The entire disclosures of the above patent applications are hereby incorporated by reference.

Field of the Invention

[0001] The present invention is concerned with a pressure sensor, a pressure control device and a pressure type flow rate control device to be employed mainly in semiconductor manufacturing facilities, chemical plants and the like, and also concerned with an automatic zero point correction device for the pressure sensor, the pressure control device and the pressure type flow rate control device which has made it possible that the occurrence of measurement errors on the pressure and flow rate due over-time to changes of pressure detection values is prevented by performing an automatic zero point correction in the event that a volume of the change (a drift volume) in the output exceeds the prescribed set value at the time when the output of a pressure sensor to measure fluid pressure changes over time.

### Background of the Invention

[0002] With semiconductor manufacturing facilities and chemical products manufacturing facilities, it is required that the flow rate and pressure of raw gases to be supplied are controlled with the high degree of accuracy. To meet the needs, various types of pressure control devices, flow rate control devices and pressure sensors to be employed for these devices have been developed.

[0003] Figure 13 and Figure 14 illustrate one example of the conventional flow rate control device. With Figure 13 (the US Patent No. 5146941), gas pressure  $P_1$  on the upstream side from the orifice F and differential pressure  $\delta P$  between the inlet side of the orifice F and the throat part are inputted to a computing means C, and the gas flow rate on the downstream side from the orifice is controlled to be the set flow rate by the open/close control of a control valve V through a valve controller VC based on the flow rate  $W_g$  computed with the said computing means and the set flow rate  $W_r$ . This is what is known as a so-called differential pressure type flow rate control device

[0004] Similarly, Figure 14 (TOKU-KAI-HEI No.8-338546) illustrates another example of the conventional pressure type flow rate control device. This is what is publicly known as a pressure type flow rate control device to be used under critical conditions ( $P_2/P_1 \leq \text{approx. } 0.5$ ), wherein the gas flow rate on the downstream side from the orifice under critical conditions is computed as  $Q_c = KP_1$  (where P is the pressure on the upstream side from the orifice) by the computing means C, and the control valve V is controlled by the open/close control to make smaller the difference between the set flow rate  $Q_s$  and the

afore-mentioned computed flow rate  $Q_c$ , thus the gas flow rate on the downstream side from the orifice F is controlled to be the set value.

[0005] With the flow rate control device and the like as described above, it is needed that the gas pressure  $P_1$  and the like on the upstream side from the orifice are detected. To detect the pressure, the pressure sensor for which semiconductor pressure sensitive elements such as a strain gauge and the like are used are widely utilized.

[0006] It has been known that, with the afore-mentioned pressure sensor to detect the fluid pressure  $P_1$ , the output values change depending on the environmental conditions surrounding the sensor, for example, such as gas temperature and the like. That is, a pressure sensor placed in the same fluid pressure might have a different output value due to the changes in fluid temperature.

[0007] For example, with the afore-mentioned strain gauge type pressure sensor, pressure is converted to voltage, and the relation that the pressure on the horizontal axis corresponds with the output voltage on the vertical axis on the graph is established. And, the output characteristics that the output voltage reaches zero when the absolute pressure is zero, and the output voltage increases linearly along with the increase of the absolute pressure are desired.

However, it has been known that, with actualthe pressure sensors in practicereality, the sensor output changes even under the same gas pressure when~~in the event that~~ gas temperature changes as described above, and that characteristics of pressure to output have no direct relation to each other in a strict sense.

[0008] ~~Specifically~~Concretely, when the pressure applied to the pressure sensor is zero, the sensor output is called a zero point output, while when the zero point changes with temperature changes, it is called a temperature drift of the zero point output, and temperature changes of the sensor output at the time of applying pressure is called a temperature drift of the span output. Adjustments on both the temperature drift of the zero point output and the temperature drift of the span output are needed to obtain an accurate sensor output.

[0009] Let's assume, for example, that the zero point voltage is 0(V) without the temperature drift of the zero point output of the pressure sensor, and that the output voltage of the pressure sensor is 20mV when the absolute pressure of  $1.0 \times 10^2$  kPaA or the gas pressure of 1 at m is applied to the pressure sensor. When the gas temperature changes under this state, it is anticipated that the output voltage changes from 20mV. As described above, the change is what is called the temperature drift of the span output. In fact, because of the temperature drift of the zero point output, what changed with the zero point voltage (the zero point output drift) are added to the temperature drift of the span output with any given pressure.

[0010] As explained above, with the pressure type flow rate control device and the like, ~~where~~while measuring the upstream side pressure  $P_1$  and/or the downstream side pressure  $P_2$ , a flow rate is controlled when passing through an orifice, there are included errors with the pressure  $P_1$ ,  $P_2$  when the output voltage is directly converted to pressure due to the reason that the temperature change characteristics which are a temperature drift of the zero point output and

the temperature drift of the span output are included in the output voltage of the pressure sensor.

[0011] For this reason, inventors of the present invention have developed system technologies which allow more accurate fluid pressure control, pressure control and flow rate control by automatically correcting the temperature drift of the zero point output and/or the temperature drift of the span output of the pressure sensor caused by the afore-mentioned temperature changes with the control circuits or control software, and made them public in the TOKU-GAN No.2001-399910.

[0012] Techniques pertaining to the afore-mentioned the TOKU-GAN No.2001-399910 make it possible to almost completely~~thoroughly~~ to eliminate control errors on the pressure, flow rate or the like arising from such a temperature drift of the pressure sensor by employing a comparatively simply constituted device, and thereby achieve excellent, practical effects.

[0013] However, it has been recently learned that there are existed not only the output voltage changes caused by the afore-mentioned fluid temperature, but also the output voltage changes over time with a pressure sensor, particularly with the pressure sensor which employs a semiconductor pressure sensitive element.

The afore-mentioned changes of the output voltage of the pressure sensor over time have become more noticeable when it is used in a state of low pressure (for example, in a vacuum of  $10^{-4} \sim 10^{-6}$  Torr to approx. 100 Torr) on the secondary side from the orifice F. Therefore, its influence has not been overlooked onwith the pressure type flow rate control device thatwhich is used

for the devices to supply various gases to the process chamber in semiconductor manufacturing facilities.

[0014] On the other hand, to eliminate effects due to the afore-mentioned output changes of the pressure sensor over time, it may be possible to formulate a measure wherein with characteristics of pressure to output of the pressure sensor are varied for a prescribed volume by installing an additional control circuit or a control software. However, this measure creates it becomes a problem because the additional installation of the device to correct these output changes over time (hereinafter called a "time-varying output drift of the pressure sensor") invites a rise in manufacturing costs of a pressure control device or a flow rate control device.

Patent Literature 1: US Patent No.5146941

Patent Literature 2: TOKU-KAI-HEI No.8-338546 Public Bulletin

Patent Literature 3: TOKU-KAI-HEI No.10-82707 Public Bulletin

## Disclosure of the Invention

### Object of the Invention

[0015] It is an object of the present invention to solve the afore-mentioned problems with the conventional pressure sensor for which a semiconductor pressure sensitive element is employed or the flow rate/pressure control device for which the pressure sensor is used, namely, (1) that control accuracy of the flow rate, pressure and the like are deteriorated due to changes of the pressure-output characteristics of the pressure sensor over time, and (2) that an additional installation of the device to correct the afore-mentioned time-varying

output drift invites a rise in manufacturing costs of the pressure control device, a flow rate control device and the like, and In accordance with the present invention, another object is to provide an automatic zero point adjustment device for a pressure sensor, a pressure control device and a pressure type flow rate control device ~~that~~which makes it possible ~~for~~that the time-varying zero point drift of the pressure sensor to be simply and accurately corrected without inviting a substantial rise in production costs by making effective use of a temperature drift correction means for correcting of the pressure-output characteristics of the pressure sensor equipped with the flow rate/pressure control device.

#### Means to Achieve the Object

[0016] Inventors of the present invention repeated various kinds of tests as shown below using not only a pressure sensor but also a pressure control device and a pressure type flow rate control device for which the sensor is employed to analyze pressure -output changes of the pressure sensor over time.

With these test results, it has been learned that, for awith the pressure sensor ~~that~~which employs a semiconductor pressure sensitive element, (1) the zero point of the pressure sensor changes over time, (2) changes of the zero point over time shift toward the minus side without exception when used under vacuum (that is, the output value under pressure zero in the pressure-output characteristics shifts toward the minus side), and (3) the zero point of the pressure sensor shifts toward the minus side, and errors of pressure control accuracy shift toward the plus side for what shifted toward the minus side (that is, if the output value shifts from the zero point toward the minus side under

pressure zero, for example, toward the minus side by voltage  $\Delta v$  equivalent to 0.2% of the full scale output voltage, errors of pressure control accuracy ~~increase~~rise by voltage  $\Delta v$  equivalent to 0.2% of the full scale output voltage).

[0017] The present invention has been created based on the knowledge obtained from what the inventors have learned as discussed above. The present invention, in accordance with a first embodiment, ~~as claimed in Claim 1~~ is fundamentally ~~so~~-constituted ~~that~~, with a pressure sensor to measure fluid pressure, the output voltage from the pressure sensor is outputted to the outside, the afore-mentioned sensor output voltage is inputted to the time-varying zero point drift correction means of the pressure sensor, a judgment is made to determine if the afore-mentioned sensor output voltage is larger than the set value with the ~~said~~-sensor output judgment means of the time-varying zero point drift correction means, and further the operating conditions of the pressure sensor are judged with the afore-mentioned operating condition judgment means of the time-varying zero point drift correction means, the time-varying zero point drift of the pressure sensor is cancelled when it is found that the afore-mentioned sensor output voltage is larger than the set value and the operating conditions of the pressure sensor are under the operating conditions previously set.

[0018] The present invention, in accordance with a second embodiment that further modifies the first embodiment, ~~as claimed in Claim 2 according to Claim 1~~ is fundamentally ~~so~~-constituted such that a semiconductor pressure sensitive element is employed as the ~~for~~-a pressure sensor, the output voltage from the pressure sensor is outputted to the outside through the amplifier and is inputted



to the time-varying zero point drift correction means of the pressure sensor through an A/D converter, and further the output for the zero point correction, which is identical to the afore-mentioned sensor output voltage and with reversed polarity, is inputted to the offset terminal of the afore-mentioned amplifier from the afore-mentioned time-varying zero point drift correction means through the D/A converter when the sensor output voltage is larger than the set value and the pressure sensor is under the set operating conditions.

[0019] The present invention, in accordance with a third embodiment, ~~as claimed in Claim 3~~ is fundamentally ~~so constituted that,~~ with ~~at~~ the pressure control device equipped with a control valve for pressure control and a pressure sensor to measure fluid pressure, the output voltage from the pressure sensor is outputted to the outside, the afore-mentioned sensor output voltage is inputted to the time-varying zero point drift correction means of the pressure sensor, a judgment is made to determine if the afore-mentioned sensor output voltage is larger than the set value with the said sensor output judgment means of the time-varying zero point drift correction means, and further the operating conditions of the pressure sensor are judged with the afore-mentioned operating condition judgment means of the time-varying zero point correction means, the time-varying zero point drift of the pressure sensor is cancelled when it is found that the afore-mentioned sensor output voltage is larger than the set value and the operating conditions of the pressure sensor are under the operating conditions previously set.

[0020] The present invention, in accordance with a fourth embodiment that further modifies the third embodiment, ~~as claimed in Claim 4 according to Claim~~

3 is fundamentally ~~so-constituted~~ such that a semiconductor pressure sensitive element is employed as the~~for~~ a pressure sensor, the output voltage from the pressure sensor is outputted to the outside through the amplifier and is inputted to the time-varying zero point drift correction means of the pressure sensor through an A/D converter, and further the output for the zero point correction, which is identical to the afore-mentioned sensor output voltage and with reversed polarity, is inputted to the offset terminal of the afore-mentioned amplifier from the afore-mentioned time-varying zero point drift correction means through the D/A converter when the sensor output voltage is larger than the set value and the pressure sensor is under the set operating conditions.

[0021] The present invention, in accordance with a fifth embodiment,~~as claimed in Claim 5~~ is fundamentally ~~so-constituted that~~, with at~~the~~ pressure type flow rate control device comprising an orifice for the flow rate control, a control valve mounted on the upstream side pipe from the orifice, and an upstream side pressure sensor installed between the orifice and the control valve to detect upstream side pressure  $P_1$  to control the flow rate of fluid passing through the orifice by the upstream side pressure  $P_1$ , the afore-mentioned output voltage from the pressure sensor is outputted to the flow rate computing means, the afore-mentioned sensor output voltage is inputted to the time-varying zero point drift correction means of the pressure sensor, a judgment is made to determine if the afore-mentioned sensor output voltage is larger than the set value with the ~~said-sensor output judgment means~~ of the time-varying zero point drift correction means, and further the operating conditions of the pressure sensor are judged with the afore-mentioned operating condition judgment means of the

time-varying zero point correction means, the time-varying zero point drift of the pressure sensor is cancelled when it is found that the afore-mentioned sensor output voltage is larger than the set value and the operating conditions of the pressure sensor are under the operating conditions previously set.

[0022] The present invention, in accordance with a sixth embodiment that further modifies the fifth embodiment, as claimed in Claim 6 according to Claim 5 is fundamentally so-constituted such that a semiconductor pressure sensitive element is employed as the for a pressure sensor, the output voltage from the pressure sensor is outputted to the outside through the amplifier and is inputted to the time-varying zero point drift correction means of the pressure sensor through an A/D converter, and further the output for the zero point correction, which is identical to the afore-mentioned sensor output voltage and with reversed polarity, is inputted to the offset terminal of the afore-mentioned amplifier from the afore-mentioned time-varying zero point drift correction means through the D/A converter when the sensor output voltage is larger than the set value and the pressure sensor is under the set operating conditions.

[0023] The present invention, in accordance with a seventh embodiment, as claimed in Claim 7 is fundamentally so-constituted that, with a the pressure type flow rate control device comprising an orifice for the flow rate control, a control valve mounted on the upstream side pipe from the orifice, an upstream side pressure sensor installed between the orifice and the control valve to detect upstream side pressure  $P_1$ , and a downstream side pressure sensor mounted on the downstream side pipe to detect downstream side pressure  $P_2$  to control the flow rate of fluid passing through the orifice by both upstream side pressure  $P_1$

and downstream side pressure  $P_2$ , the output voltage from the pressure sensor is outputted to the flow rate computing means, the afore-mentioned sensor output voltage is inputted to the time-varying zero point drift correction means of the pressure sensor, a judgment is made to determine if the afore-mentioned sensor output voltage is larger than the set value with the said sensor output judgment means of the time-varying zero-point drift correction means, and further the operating conditions of the pressure sensor are judged with the afore-mentioned operating condition judgment means, the time-varying zero point drift of the pressure sensor is canceled when it is found that the afore-mentioned sensor output voltage is larger than the set value and the operating conditions of the pressure sensor are under the operating conditions previously set.

[0024] The present invention, in accordance with an eighth embodiment, ~~as claimed in Claim 8 according to Claim 7~~ is fundamentally ~~so-constituted~~ such that a semiconductor pressure sensitive element is employed as the ~~for a~~ pressure sensor, the output voltage from the pressure sensor is outputted to the outside through the amplifier and is inputted to the time-varying zero point drift correction means of the pressure sensor through an A/D converter, and further the output for the zero point correction, which is identical to the afore-mentioned sensor output voltage and with reversed polarity, is inputted to the offset terminal of the afore-mentioned amplifier from the afore-mentioned time-varying zero-point drift correction means through the D/A converter when the sensor output voltage is larger than the set value and the pressure sensor is under the set operating conditions

[0025] The present invention, in accordance with a ninth embodiment that further modifies the third and fourth embodiments, as claimed in Claim 9 according to Claim 3 or Claim 4 is so made that the set value, which serves as a reference at the sensor output judgment means of the time-varying zero point drift correction means of the pressure sensor<sub>1</sub>, becomes the sensor output voltage equivalent to less than control accuracy of the full scale pressure FS to be detected by the pressure sensor.

[0026] The present invention, in accordance with a tenth embodiment that further modifies the third and fourth embodiments, as claimed in Claim 10 according to Claim 3 or Claim 4 is so made that the set operating conditions, which serve as a reference at the operating condition judgment means of the time-varying zero point drift correction means of the pressure sensor<sub>1</sub>, are made up of three conditions, that is, whether or not a signal for forced opening to the control valve exists, whether or not a signal for forced closing to the control valve exists, and the set signal for the flow rate is zero.

[0027] The present invention, in accordance with an eleventh embodiment of the invention that further modifies the fifth, sixth, seventh and eighth embodiments, as claimed in Claim 11 according to Claim 5, Claim 6, Claim 7 or Claim 8 is so-made such that the set value, which serves as a reference at the sensor output judgment means of the time-varying zero point drift correction means of the pressure sensor<sub>1</sub>, becomes the sensor output voltage equivalent to less than control accuracy of the full scale pressure FS to be detected by the pressure sensor.

[0028] The present invention, in accordance with a twelfth embodiment that

further modifies the fifth, sixth, seventh and eighth embodiments, as claimed in Claim 12 according to Claim 5, Claim 6, Claim 7 or Claim 8 is so made so that the set operating conditions, which serve as a reference at the operating condition judgment means of the pressure sensor, are made up of three conditions, that is, whether or not a signal for forced opening to the control valve exists, whether or not a signal for forced closing to the control valve exists, and the set value of the flow rate is zero.

[0029] The present invention, in accordance with a thirteenth embodiment that further modifies the fourth embodiment, as claimed in Claim 13 according to Claim 4 is so made such that a D/A converter, through which voltage for the zero point correction is outputted to the offset terminal of the amplifier from the time-varying zero point correction means, is shared with the temperature drift correction means of the pressure sensor mounted on the flow rate computing means of the said-pressure type flow rate control device.

[0030] The present invention, in accordance with a fourteenth embodiment that further modifies the sixth embodiment or the eighth embodiment, as claimed in Claim 14 according to Claim 6 or Claim 8 is so constituted such that a D/A converter, through which voltage for the zero point correction is outputted to the offset terminal of the amplifier from the time-varying zero point drift correction means, is shared with the temperature drift correction means of the pressure sensor mounted on the flow rate computing means of the said-pressure type flow rate control device.

### Effects of the Invention

[0031] ~~T~~According to the present invention, in accordance with the first embodiment as claimed in Claim 1, it is so constituted that is constituted so the zero point drift ~~that~~which occurs with changes over-time is cancelled based on the judgment by the time-varying zero point drift correction means, thus resulting in substantial improvements in pressure detection accuracy of the pressure sensor.

[0032] ~~T~~According to the present invention, in accordance with the second embodiment as claimed in Claim 2, it is so constituted so that the zero point drift ~~that~~which occurs with changes over time is cancelled by inputting voltage, which is identical to the drift voltage occurred with changes over time of the pressure sensor and with reversed polarity, to the offset terminal of an amplifier for amplifying output of the pressure sensor based on the judgment by the time-varying zero point drift correction means, thus resulting in substantial improvements in pressure detection accuracy of the pressure sensor.

[0033] Also according to the third to eighth embodiments, inclusive, of the present invention as claimed in Claim 3 to Claim 8 inclusive, control accuracy of pressure and flow rate is remarkably improved ~~because~~due to the reason that pressure detection accuracy of the pressure sensor, which functions as the basis of pressure control and flow rate control, is enhanced.

[0034] According to the ninth and eleventh embodiments of the present invention as claimed in Claim 9 and Claim 11, an automatic zero point correction is conducted based on the sensor output voltage less than control accuracy of the full scale pressure FS to be detected by the pressure sensor, for example,

that is equivalent to 0.13% of the full scale FS, thus always allowing the flow rate measurement value to be maintained within the prescribed range of accuracy.

[0035] Furthermore, according to the tenth and twelfth embodiments of the present invention~~-as claimed in Claim 10 and Claim 12~~, the zero point drift can be cancelled with athe high degree of accuracy because~~due to the reason that~~ the time-varying zero point drift correction is automatically performed in the event that the environmental conditions surrounding the pressure sensor on the upstream side from the orifice are in a state of ~~the~~ near-vacuum.

[0036] According to the thirteenth and fourteenth embodiments of the present invention~~-as claimed in Claim 13 and Claim 14~~, a D/A converter to supply voltage for the drift correction to the offset terminal of an amplifier is made in such a manner that it is shared with the temperature drift correction means of the pressure sensor, thus allowing the constitutions of the correction means of the temperature drift and the time-varying drift of the pressure sensor for the pressure control device and the pressure type flow rate control device to be simplified. As stated above, the present invention achieves excellent, practical effects.

#### Brief Description of the Drawings

[0037] Figure 1 is a structural drawing of a semiconductor element type pressure sensor (a pressure transducer) used for the present invention.

Figure 2 is a cross-sectional view to illustrate how the pressure sensor used for the present invention is mounted.

Figure 3 is a curve to show changes over-time of the zero point output



of the pressure sensor used for the present invention while being maintained under vacuum.

Figure 4 is a diagram to show "the difference by the service history before the evacuation" of changes over-time of the zero point output of the pressure sensor used for the present invention while being maintained under vacuum.

Figure 5 is a diagram to show changes over-time of the zero point output in the event that pressure of the pressure sensor used for the present invention is made to change at a cycle of 0 (Torr · vacuum)~60Torr.

Figure 6 is a diagram to show changes over-time of the zero point output in the event that pressure of the pressure sensor is made to change at a cycle of 0(Torr · vacuum)~0.1MpaG.

Figure 7 is a diagram to show changes over-time of the zero point output in the event that pressure of the pressure sensor is maintained at 0.1MpaG.

Figure 8 is a block diagram of the pressure type flow rate control device used for the ~~embodiments~~ of the present invention.

Figure 9 is a block diagram of the zero point output correction part in the means to correct the output changes of the pressure sensor of the pressure type flow rate control device by temperature used for the ~~embodiments~~ of the present invention.

Figure 10 is an explanatory drawing to show the relationship between the correction of the zero point output voltage and the full scale (FS) of the pressure sensor.

Figure 11 is a block diagram of the control circuit of the pressure type flow rate control device pertaining to the present invention.

Figure 12 is an operating flow diagram of the time-varying zero-point drift correction means of the pressure sensor pertaining to the present invention.

Figure 13 shows one example of a conventional pressure type flow rate control device.

Figure 14 shows another example of a conventional pressure type flow rate control device.

#### List of Reference Characters and Numerals

[0038] P1 Gas pressure on the upstream side from the orifice

A Pressure sensor

B Pipe

1 Pressure type flow rate control device

2 Orifice

2a Orifice opening

3 Upstream side pressure sensor

4 Upstream side pipe

5 Downstream side pipe

6 Temperature sensor

7 Control circuit

7a Flow rate computing means

7b Flow rate setting means

7c Comparative means

- 8 Valve driving part
- 9 Control valve
- 10 Gas supply source
- 11 Pressure regulator
- 12,13 Valves
- 14 Process chamber
- 15 Vacuum pump
- 16 Fixed amplification circuit
- 16a Offset terminal
- 17,18 Variable amplification circuits
- 19 A/D converter
- 20 CPU
- 21 Sensor base
- 22 Sensor chip
- 23 Diaphragm
- 24 Diaphragm base
- 25 Silicon oil
- 26 Sealing body
- 27 Lead pin
- 28 Fixing body
- 29 Holding nut
- 30 Bearing
- 31 Seal ring
- 40 D/A converter for offsetting

40a,40b D/A converters  
40c,40d Buffers  
40e Buffer for synthesis  
41 D/A converter  
42,43,44 A/D converter  
7a' Flow rate linearity correction part in the flow rate computing means  
48 Temperature drift correction means of the pressure sensor  
49 Time-varying zero point drift correction means of the pressure  
sensor  
49a Sensor output judgment means  
49b Operating condition judgment means  
50 Piezo-step-up circuit

#### Best Mode to Carry Out the Invention

[0039] First, inventors of the present invention conducted a survey and then measured changes over-time of pressure-output characteristics of a pressure sensor A by fixing the pressure sensor A structured as shown in Figure 1 to a pipe B in a configuration as shown in Figure 2, and by the inside of the pipe B being evacuated and maintaining in the prescribed degree of vacuum by use of a vacuum pump (not illustrated).

Referring to Figure 1 and Figure 2, 21 designates a sensor base, 22 a sensor chip (a semiconductor type pressure sensitive element), 23 a diaphragm, 24 a diaphragm base, 25 silicon oil, 26 a sealing body, 27 a lead pin, 28 a fixing body, 29 a holding nut, 30 a bearing, 31 a seal ring and P<sub>1</sub> gas pressure.

[0040] Referring to Figure 2, a pressure sensor A is fixed to the fixing body 28 by use of the holding nut 29. However, any fixing mechanisms can be employed to fix the pressure sensor A. For example, a flange (not illustrated) for fixing can be used to fix the pressure sensor A to the fixing body 28.

Though not illustrated in Figure 1 and Figure 2, a pressure sensor A having a so-called strain gauge fixed to the inside surface of the diaphragm 23 and having a structure wherein silicon oil 25 is not used has been also used to replace the pressure sensor A having a structure shown in Figure 1.

[0041] Gas pressure  $P_1$  applied to a diaphragm 23 changes by decompressing the inside of the pipe B, and accordingly pressure applied to the sensor chip 22 (or a strain gauge) changes, thus resulting in changes in the output voltage from the sensor chip 22, and changes in gas pressure  $P_1$  is detected. Because the pressure sensor A itself is publicly disclosed (the TOKU-KAI-HEI No.10-82707 and others), the explanation is omitted herewith.

[0042] Figure 3 is a diagram to show a state of changes in the zero point of the pressure sensor A in the event that the pressure sensor A is fixed as shown in Figure 2, and maintained under vacuum (the degree of vacuum  $10^{-5} \sim 10^{-6}$  Torr) after leaving it alone for 24 hours under atmospheric pressure.

As apparent from Figure 3, in approximately one hour immediately after the start of evacuation, zero point changes toward the minus direction by 0.2~0.3%FS (changes are made for 0.2~0.3Torr when the full scale FS is made to be 100 Torr), and in approximately 5 hours later, it changes further toward the minus direction by 0.1%FS, and then, without it being stabilized, it keeps changing toward the minus direction although the volume of changes is small.

The output of the pressure sensor on the vertical axis in Figure 3 is expressed by mV, and 2mV is equivalent to 0.1%FS (that is, 0~100 Torr is equivalent to the output voltage 0~2V).

[0043] Figure 4 shows how pressure · time, which the pressure sensor experiences before evacuation, influences over the stabilization time of the zero point. That is, a test specimen, which is found to be in a state of the zero point being stable to some degree by the vacuum maintenance test, is given experiences in having a couple of different pressure, and stabilization time of the zero point is continuously monitored by maintaining it under vacuum to investigate effects that pressure experienced before evacuation exerted on the stabilization time of the zero point thereafter.

[0044] As apparent from Figure 4, the higher pressure experienced is, the higher the initial value of the zero point is, and the degree of changes toward the minus side after evacuation becomes larger. However, after the lapse of 20~30 hours, it settles down to nearly the same value irrespective of pressure experienced previously. Then, after that, similar to the result of the measurement test on stabilization time of the zero point when left alone in vacuum in Figure 3, it is assumed that it keeps decreasing at a certain percentage. Time shown by an arrow in the figure is time elapsed from the initial time of the continuous monitoring.

[0045] Figure 5 and Figure 6 show the results of measurement wherein the pressure range of 0 Torr (held for 15 seconds and with the vacuum degree of approximately  $10^{-6}$  Torr)~60Torr (held for 30 seconds) is changed with regularity and continuously over 5 hours per day (Figure 5), and also the pressure range of

0 Torr (held for 15 seconds and with the vacuum degree of approximately  $10^{-6}$  Torr)~0.1-1MPaG (held for 30 seconds) is changed with regularity and continuously over 5 hours per day (Figure 6), thus finding the changes in the zero point respectively after the lapse of 5 hours. The measurement of the changes in the zero point was conducted over 4 weeks at the interval of approximately one week.

[0046] Apparent from Figure 5, it is found that there occurred some changes in the zero point of 0.2%FS over 1-2 weeks.

With the result in Figure 6, it is also found that changes in the zero point remain within 0.1%FS toward the minus direction.

[0047] Figure 7 shows the changes over-time in the zero point output in the event that the pressure sensor A is continuously held under pressurization of 0.1MPaG. It is found that changes in the zero point occurs with the volume of changes less than 0.1%FS toward the plus side.

[0048] From the test results shown in Figure 3 to Figure7 inclusive, it is found that there existed the below-stated events on the changes over-time in the zero point output with the pressure sensor A which employs a semiconductor pressure sensitive element (a transducer).

[0049]

1. With the cycle test of vacuum maintaining, vacuum $\leftrightarrow$ 60Torr, the zero point keeps changing toward the minus direction.
2. With a vacuum maintaineding, prominent changes are shown in the initial few hours.
3. Though the percentage of change decreases with the lapse of time, it is found

that relatively large changes are seen in the initial stage when it is returned to atmospheric pressure or it is evacuated after it was put in a state of being pressurized to 0.1MPaG.

4. Unevenness is seen in the vacuum⇔60Torr cycle test. With some, it is found that the volume of changes is larger than that in the test of vacuum maintaining. After the lapse of one week, it is found that some are out of 0.2%FS.

5. It is assumed that from the results of 0.1MPaG maintaining test, the zero point does not change with this much maintained pressurization-maintaining. It is also understood that there are not big changes when it is in a state of atmospheric pressure.

[0050] Based on the changes over-time in the zero point output of the pressure sensor A as stated above, inventors of the present invention have created a measure to automatically correct the over-time changes in the zero point not only of the pressure sensor A, but also of the pressure control device and the pressure type flow rate control device which employ the pressure sensor A.

[0051] An automatic zero point adjustment device to correct changes over-time in the zero point output of the pressure type flow rate control device according to the present invention is explained hereafter with reference to the drawings.

[0052] Figure 8 is a block diagram of the pressure type flow rate control device by making use of critical conditions according to the present invention. With the pressure type flow rate control device 1, the flow rate is expressed with  $Q=KP_1$  and the measurement of pressure is conducted only by the upstream side pressure sensor 3 due to the reason that it is based on the premise that fluids to be supplied are under critical conditions, that is, the flow velocity of fluids flowed



out through the orifice 2 is that of sound. The pressure type flow rate control device 1 is equipped with an orifice 2 which forms an orifice 2a, an upstream side pipe 4, a downstream side pipe 5, an upstream side pressure sensor 3, a temperature sensor 6, a control circuit 7, a valve driving part 8 and a control valve 9.

[0053] A control circuit 7, which is constituted centering around an electronic circuit, a micro-computer and a built-in program, comprises an electronic circuit system such as an amplification circuit not illustrated, an A/D converter and the like, a flow rate computing means 7a to compute a flow rate  $Q_c$  using the experimental flow rate equation, a flow rate setting means 7b to command the set flow rate  $Q_s$  to flow, and a comparative means 7c to compute the flow rate difference  $\Delta Q (=Q_s - Q_c \text{ or } Q_c - Q_s)$  between the flow rate  $Q_c$  to be computed and the set flow rate  $Q_s$ .

[0054] Referring to Figure 8, 10 designates a gas supply source, 11 a pressure regulator, 12,13 valves, 14 a process chamber and 15 a vacuum pump.

[0055] Pressure  $P_2$  on the downstream side from the orifice 2 is set considerably smaller than pressure  $P_1$  on the upstream side by evacuation using a vacuum pump, automatically to hold the critical condition of  $P_2/P_1 < \text{approximately } 0.5$  at any time. As a result, the velocity of the gas to flow out through the orifice opening becomes that of sound, and the flow rate  $Q$  passing through the orifice 2 is expressed as  $Q=KP_1$ .

[0056] Pressure  $P_1$  on the upstream side is measured by the pressure sensor 3. For the accurate measurement of pressure, the sensor part of the pressure sensor 3 is brought into contact with the gas flow. Furthermore, the sensor part

is designed to be minimal so that the gas flow can stay out of turbulence. Accordingly, temperature of the sensor part becomes equal to gas temperature T.

Gas temperature T is measured by the temperature sensor 6. Temperature in the proximity of the orifice 2 is measured by the temperature sensor 6 so that the gas flow can stay out of turbulence. When both the gas and the orifice reach a state of equilibrium, thermally, their temperature becomes equal. Therefore, temperature of the orifice can be measured as that of the gas.

[0057] Pressure  $P_1$  on the upstream side and temperature T of the gas are provided as voltage, and ~~converted~~ turned to be digital signals by the amplification circuit or the A/D converter. The digital signals are inputted to the flow rate computing means 7a, and proportional coefficient K is computed from the gas temperature

T and gas properties, and the flow rate  $Q_c$  to be computed by the equation  $Q_c = KP_1$  by making use of pressure  $P_1$  on the upstream side.

The set flow rate  $Q_s$  has been inputted from the flow rate setting means 7b, and thus, the flow rate  $\Delta Q$  is computed with equation  $\Delta Q = Q_s - Q_c$  by the comparative means 7c.

Computed flow rate difference  $\Delta Q$  is outputted to the valve driving part 8, and the degree of opening of the control valve 9 is adjusted to make  $\Delta Q$  zero. With the opening adjustment, pressure  $P_1$  on the upstream side from the orifice is variably adjusted, thus the flow rate  $Q_c$  to be computed and obtained with the equation  $Q_c = KP_1$  being controlled to be equal to the set flow rate.

[0058] As stated above, temperature of the sensor part of the pressure sensor 3 is made to be equal to gas temperature  $T$ , and temperature of the sensor part of the pressure sensor changes as gas temperature changes. And, the pressure sensor 3 has temperature dependency so that the output voltage of the pressure sensor changes as temperature changes. Accordingly, the pressure type flow rate control device according to the present invention is equipped with a device to correct changes (drift) of the output voltage caused by temperature with the pressure sensor 3 as shown in Figure 9.

[0059] Figure 9 is a simplified block circuit diagram of the temperature drift correction device to be used for adjustment of the zero point output (that is, output voltage under a state of pressure being zero) in the means to correct changes (drift) of output voltage caused by the afore-mentioned temperature with the pressure type flow rate control device.

Referring to Figure 9, output voltage  $V$  of the pressure sensor 3 is amplified up to pressure voltage  $V$  by the fixed amplification circuit 16 and the variable amplification circuit 18. Pressure voltage  $V$  is inputted to the CPU 20 through the mediation of the A/D converter 19. Output of the fixed amplification circuit 16 is outputted to the other variable amplification circuit 17, and output of the variable amplification circuit 17, which is output voltage, is displayed on the digital panel as pressure  $P_1$  on the upstream side.

[0060] Assuming that 100 mV is outputted at the time when the afore-mentioned pressure sensor 3 perceives absolute pressure  $P_1 = 7$  atmospheric pressure (that is,  $7 \times 10^2$  kPaA), output voltage  $V$  of the pressure sensor 3 becomes output voltage in the range of  $V = 0 \sim 42.86$  V when pressure

$P_1$  on the upstream side is controlled in the range of  $P_1 = 0 \sim 3(x10^2 \text{ kPaA})$  by the pressure sensor.

If the maximum voltage 42.86 mV of output voltage V is amplified to the full scale of 5V, the amplification rate becomes 117 times. With this embodiment, a 117-times amplification rate has been realized by amplifying making 100 times with the afore-mentioned fixed amplifier 16 and 1.17 times with the variable amplifiers 17, 18.

[0061] Output voltage of the pressure sensor 3 drifts with temperature changes. Output changes (drift) at zero pressure is called a zero point output temperature drift, and output changes (drift) under any given pressure is called an output temperature drift.

The afore-mentioned zero point output temperature drift is corrected by adjusting the offset terminal 16a of the fixed amplifier 16. SpecificallyConcretely, correction of the zero point output drift is realized by the D/A converter for offsetting. That is, when output voltage V indicates a certain value  $+v_0$  at the time of pressure zero,  $-v_0$  is inputted to the offset terminal 16a, to make the zero point output drift voltage zero. As a result, changes (drift) in the zero point output have been corrected as effective input voltage becomes  $v_0 + (-v_0) = 0$  even when output voltage  $v_0$  is inputted to the fixed amplifier 16 at pressure zero.

[0062] The afore-mentioned D/A converter for offsetting comprises a D/A converter 40a for coarse adjustment and a buffer 40c, a D/A converter 40b for fine adjustment and a buffer 40d, and a buffer 40e for synthesis. As explained as followsherewith, the zero point output drift is corrected by canceling the zero

point output drift by means of applying the zero point correction voltage  $-v_0$ , for which the zero point output drift voltage  $v_0$  is reversed, to the offset terminal 16a by the circuit for coarse adjustment and the circuit for fine adjustment.

[0063] Figure 10 is an explanatory drawing to show the relationship between the correction of changes (drift) in the zero point output voltage and the setting of the full scale FS. Pressure  $P_1$  on the upstream side is shown on the lateral axis, and output voltage  $v$  of the pressure sensor 3 and pressure voltage  $V$  of the variable amplifier 18 is shown on the vertical axis. The pressure range is made to be  $P_1=0\sim P_{1m}$ , and the maximum pressure is made to be  $P_{1m} = 3.0(\times 10^2 \text{ kPaA})$ . Now, when gas temperature  $T$  is  $T_0$ , the maximum output voltage of the sensor is  $v_1 = 40.8 \text{ mV}$  at the zero point drift of  $v_0 = -2.0 \text{ mV}$  and the maximum pressure  $P_{1m}$ .

[0064] Thus, a dotted line  $a'$  connecting  $v_0$  and  $v_1$  in Figure 10 shows the temperature characteristics of the pressure sensor 3. Here, when  $-v_0$  is applied to the offset terminal 16a,  $v_0$  becomes  $0 \text{ mV}$  with the equation  $v_0 + (-v_0) = 0$ , thus making the correction from  $v_0$  to zero (an arrow  $a$ ). As the result, the sensor output voltage at the maximum pressure  $P_{1m}$  also becomes  $v_0 + (-v_0) = 40.8 + 2.0 = 42.8 \text{ mV}$ . Accordingly, the output of the pressure sensor 3 is corrected to  $0\sim 42.8 \text{ mV}$  with the zero point drift correction. The temperature characteristics after the correction is shown by the dashed line  $a''$ .

[0065] Next, the full scale setting of the pressure sensor 3 is performed. When the output of the pressure sensor after the zero point adjustment is  $0\sim v_1 + (-v_0)$ , that is,  $0\sim 43.8 \text{ mV}$ , it is set to the full scale  $5\text{V}$ . That is, to amplify  $42.8 \text{ mV}$  to  $5\text{V}$ , the amplification rate of the variable amplifiers 44, 46 is made to be

1.17. As the result, the 2-step amplification rate is set at  $M=100 \times 1.17=117$ .

This correction is shown by the arrow 6.

[0066] Accordingly, the maximum voltage  $V_m$  becomes  $V_m=(v_1-v_0)$ , and the output voltage  $v$  of the pressure sensor 3 at any given pressure  $P_1$  is amplified to  $V=M(v-v_0)$ . The solid line C represents the amplified output  $V$ . For the critical condition,  $V=a(T_0)P_1$  is represented by the solid line C. The proportional constant  $a(T_0)$  shows the proportional constant when gas temperature  $T$  is  $T_0$ .

[0067] With the explanation on the afore-shown Figure 9 and Figure 10, it is known that the zero point output drift of  $v_0=-2.0\text{mV}$  is caused by temperature changes of fluids (gases). Accordingly, it is understood that the straight line  $a'$  represents the temperature characteristics of the output  $v$  of the pressure sensor, and the straight line C represents the temperature characteristics of the output  $V$  of the amplifier.

[0068] On the other hand, with the present invention, the issue is the zero point correction of the over-time changes of the output  $v$  of the pressure sensor. Therefore, by defining that the zero point output drift  $v_0$  in the afore-shown Figure 9 and Figure 10 (that is, the output  $v_0$  of the output sensor at pressure zero) is the zero point output drift caused by the changes over-time, the correction means explained on Figure 9 and Figure 10 previously and its relation with the setting of the full scale (FS) become applicable, as it is, to the correction of the zero point output drift cause by the changes over-time.

That is, it is understood ~~can be grasped~~ that the straight line  $a'$  with the afore-shown Figure 10 is the time-varying characteristics of the output  $v$  of the pressure sensor, and that the straight line C is the time-varying characteristics of

the output V of the amplifier.

[0069] Detailed explanation on how to measure the over-time output change characteristics of the output v of the pressure sensor 3 illustrated in the afore-shown Figure 3 to Figure 6 inclusive is omitted herewith. The lapse of time and the volume of changes (drift voltage  $v_0$ ) of the zero point output of the pressure sensor 3 are measured by making the pressure inside the pipe B, wherein the pressure sensor 3 is assembled in the shape as shown in Figure 2, pressure zero (vacuum  $10^{-5} \sim 10^{-6}$  Torr), that is,  $P_1 \doteq 0(x10^{-2}$  kPaA) by use of a vacuum pump (not illustrated), or the lapse of time and the volume of changes  $v_0$  of the zero point output of the pressure sensor 3 are measured under a state of the pressure inside the pipe B being held at any given set value.

[0070] The afore-shown Figure 3 is a characteristic diagram of time-varying output which illustrates one example of the zero point output drift by the time changes of the pressure sensor 3 when maintained under vacuum. Time (Hr) is shown on the lateral axis, while voltage  $v_0$  of the zero point output drift is shown on the vertical axis. The output voltage 2mV of the sensor is equivalent to 0.1% of the full scale at the time when the full scale is made to be 100 Torr. The line of the PT output 0 mV shows an ideal situation having no drift. The curve represents the zero point output drift actually measured. The drift differs depending on samples of the pressure sensor. However, as stated above, it will be approximately 0.2~0.3%FS ( $v_0=4\sim6$  mV) after approximately 1 hour, and approximately 0.4%FS ( $v_0=8$  mV) after approximately 6 hours, and the zero point output drift voltage  $v_0$  is applied to the offset terminal 16a of the fixed amplification circuit 16 in the afore-shown Figure 9.

[0071] With the present invention, it is made in such a manner that when the time-varying zero point output drift  $v_0$  of the pressure sensor 3 becomes larger than  $-0.13\%FS$  (that is, the zero point output drift  $v_0$  of  $-2.6mV$ ), an automatic zero point adjustment of the pressure sensor 3 is performed by applying the zero point output drift  $v_0$  to the offset terminal 16a of the fixed amplification circuit 16 shown in Figure 9.

The reasons why the afore-mentioned  $-0.13\%FS$  is made to be an adjusting reference point for the time-varying zero point drift are that, after the result of the basic tests shown in Figure 3 to Figure 6 and others, it has been learned that the zero point drift  $v_0$  can occur only toward the minus direction when maintained under vacuum, and also that the zero point drift of the order of  $-0.13\%FS$  ( $v_0 = -2.6mV$ ) is within practical tolerance of the pressure sensor 3.

[0072] Specifically ~~Concretely~~, first, a judgment is made to see whether or not the output voltage  $v$  of the pressure sensor 3 is on the minus side.

It should be noted that gas pressure has always been applied to the pressure sensor 3 while a pressure control device is in use. This means that there is no chance that the output voltage  $v$  of the pressure sensor 3 is on the minus side. Accordingly, when it is judged that the output voltage  $v$  of the pressure sensor 3 is on the minus side, it is known that the pressure control device is not in use, and that no gas is running.

[0073] When maintained under vacuum, it is certain that the time-varying zero point output voltage drift of the pressure sensor 3 is always on the minus side. Therefore, it is understood that the pressure sensor 3 is maintained under vacuum or near-vacuum (of the order of  $10^{-2} \sim 10^{-6}$  Torr) if the output drift  $v$  of the



pressure sensor 3 is found on the minus side.

[0074] Accordingly, by judging that the output voltage drift  $v$  of the pressure sensor 3 is on the minus side, the adjustment of the time-varying zero point drift can be performed anytime because it tells that the pressure type flow rate control device is in a state of no use, and the pressure inside the pipe is maintained in near-vacuum.

[0075] Next, a judgment is made to find if the output voltage drift  $v$  of the pressure sensor 3 exceeds the afore-mentioned set value ( $v = -0.13\%FS$ ) or not. And, in case that it is found that the output drift  $v$  of the pressure sensor exceeds the set value, the adjustment of the zero point drift  $v_0$  is automatically performed with the self-test wherein it is judged that the adjustment of the time-varying zero point drift of the pressure sensor 3 is required.

[0076] Figure 11 is a detailed block diagram of the control circuit of the pressure control device. The explanation is omitted herewith because a pressure sensor 3, a fixed amplifier 16, variable amplifiers 17, 18, an A/D converter 19, a D/A converter 4 for offsetting, and the like are the same as those shown in Figure 9.

[0077] The control circuit of the pressure type flow rate control device is also nearly the same as that in Figure 11. The only difference from Figure 11 is that the gas temperature part (not illustrated) is mounted on the output side of the flow rate linearity correction part 7a' in the flow rate computing means, and the temperature detection signal from the temperature sensor 4 is inputted to the gas temperature correction part.

With Figure 11, 41 designates a D/A converter, 42, 43, 44 A/D

converters, 7 a control circuit, 7a a comparative circuit, 20 a CPU, 7a' a flow rate linearity correction part in the flow rate computing means, 48 a temperature drift correction means of the pressure sensor, 49 a time-varying zero point drift correction means of the pressure sensor, and 50 a piezo-step-up circuit. The control valve (not illustrated) is controlled for opening and closing with the output from the said-piezo-step-up circuit 50.

[0078] The time-varying zero point drift correction means 49 of the pressure sensor is equipped with a means to judge if the input value  $v$  from the A/D converter 44 exceeds the set value ( $-0.13\%FS = -2.6mV$ ) (a sensor output judgment means 49a), and an operating condition judgment means 49b to judge either if the input for forced closing to a control valve 9 is set or if the pressure setting signal  $V$  is less than  $0.6\%FS$ . When one of the following occurs, that is, (1) if input for forced opening of the control valve 9 is set, (2) if input for forced closing of the control valve 9 is set, or (3) if the pressure setting signal  $V$  is less than  $0.6\%FS$  ( $V = 60\text{ mV} \cdot \text{sensor output voltage } v = 12\text{ mV}$ ) is confirmed, and also when it is confirmed that output  $v$  of the pressure sensor 3 is more than  $-0.13\%FS$  by the sensor output judgment means 49a, then voltage for the zero point adjustment ( $v_0 = 2.6\text{ mV}$ ), which is equivalent to  $+0.13\%FS$ , is automatically inputted to the offset terminal 16a of the fixed amplification circuit 16 from the D/A converter, thus performing the automatic zero point adjustment by the drift output ( $-2.6\text{ mV}$ ) equivalent to the time-varying zero point drift ( $-0.13\%FS$ ) of the pressure sensor being canceled therewith.

[0079] Figure 12 is an operating flow diagram of the time-varying zero point drift correction means 49 of the pressure sensor according to the present

invention. With the step m<sub>1</sub>, output voltage  $v$  from the pressure sensor 3 is inputted. With the step m<sub>2</sub>, the input signal  $V_c$  for forced opening, or the input signal  $V_o$  for forced closing, is inputted to the control valve 9. With the step m<sub>3</sub>, a judgment is made to determine if the afore-mentioned  $v$  exceeds  $-0.13\%FS$  ( $v = -2.6 \text{ mV}$ ) or not. With the step m<sub>4</sub>, a judgment is made to determine if  $V_c$  or  $V_o$  exists, and also if the pressure setting signal  $V$  is less than  $0.6\%FS$ .

Lastly, with the step m<sub>5</sub>, when  $v$  exceeds  $-12\text{mV}$  and either one of the conditions  $V_c > 0$  or  $V_o > 0$ , or  $V < 0.6\%FS$  is satisfied (step m<sub>5</sub>), then output voltage of  $+v (= 2.6 \text{ mV})$  is outputted to the offset terminal 16a of the fixed amplification circuit 16 with the step m<sub>6</sub>.

[0080] With the embodiment according to the present invention illustrated in the afore-shown Figure 1 to Figure 11 inclusive, the present invention was explained based on the pressure type flow rate control device to be used under critical conditions. However, there is no need to say that the present invention can be applied to the pressure control device to be used under non-critical conditions or the pressure sensor to be used singly.

#### Feasibility of the Industrial Use

[0081] The present invention is mainly used for semiconductor manufacturing facilities or chemical products manufacturing facilities. The present invention is also widely used in the fields where the high degree of accuracy is required to control the flow rate or the supply pressure of fluids such as raw gases.

FIG. 12

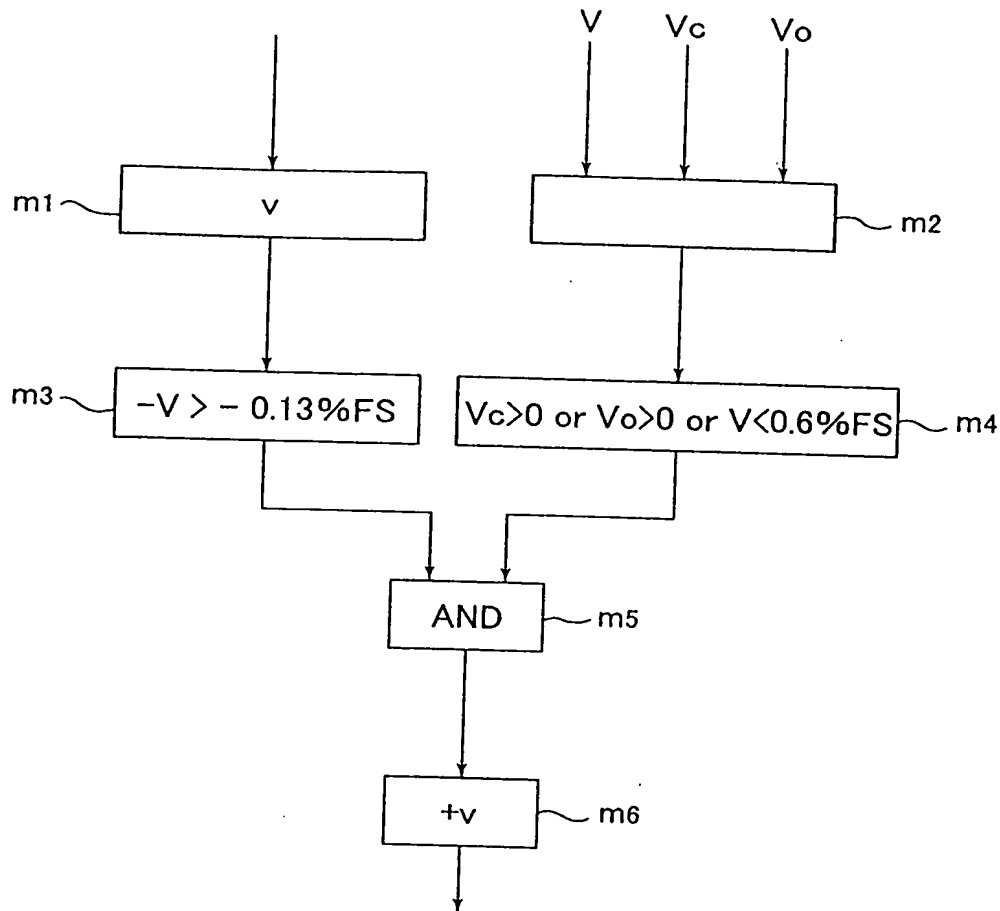


FIG. 13

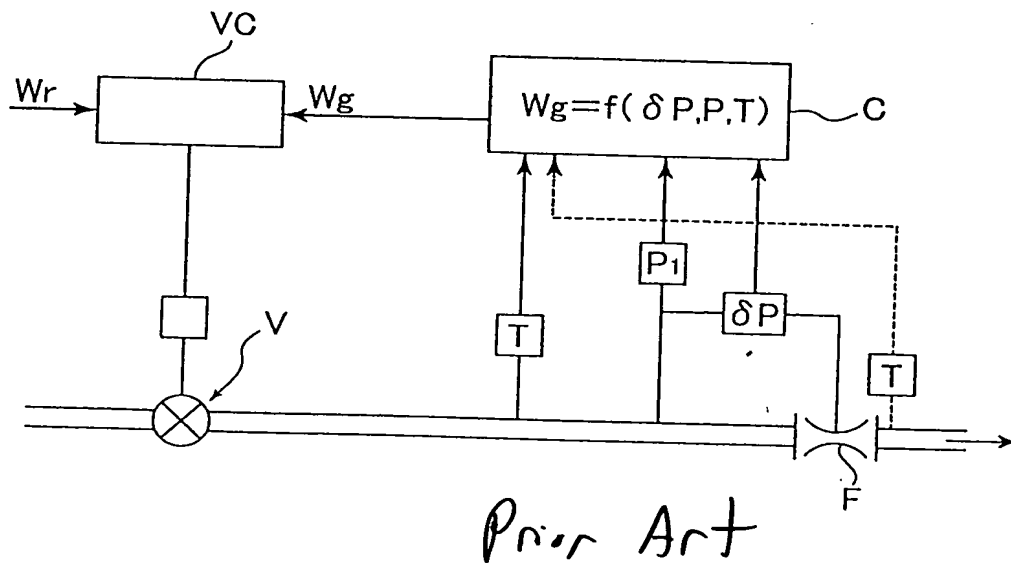
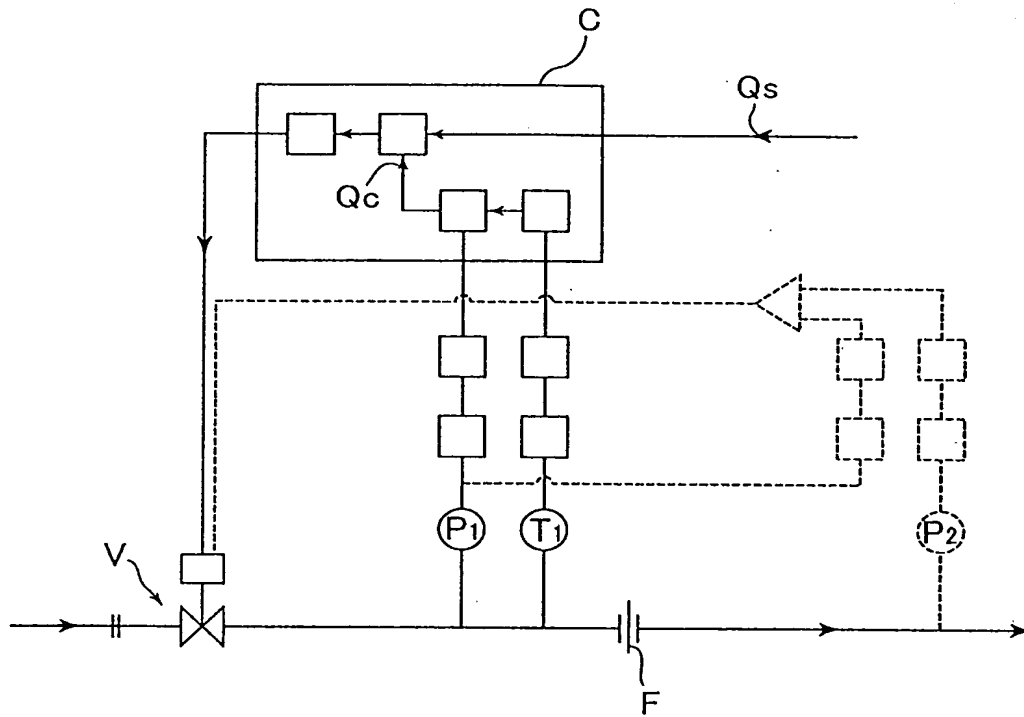


FIG. 14



Prior Art